

# (12) UK Patent Application (19) GB (11) 2 319 913 (13) A

(43) Date of A Publication 03.06.1998

(21) Application No 9803132.1

(22) Date of Filing 24.06.1994

Date Lodged 12.02.1998

(30) Priority Data

(31) 05155351 (32) 25.06.1993 (33) JP

(62) Divided from Application No 9412705.7 under Section 15(4) of the Patents Act 1977

(71) Applicant(s)

NEC Corporation

(Incorporated in Japan)

7-1, Shiba 5-Chome, Minato-Ku, Tokyo, Japan

(72) Inventor(s)

Nozomu Watanabe

(51) INT CL<sup>6</sup>

H04B 1/40, H03L 7/00

(52) UK CL (Edition P)

H3A AB AL1 AL2DX AL2D2

U1S S2204 S2205

(56) Documents Cited

None

(58) Field of Search

UK CL (Edition P) H3A AB AQA AQX AXB AXC AXD

AXE AXF AXX

INT CL<sup>6</sup> H03D, H03J, H04B

Online: WPI, JAPIO, INSPEC

(74) Agent and/or Address for Service

Mathys & Squire

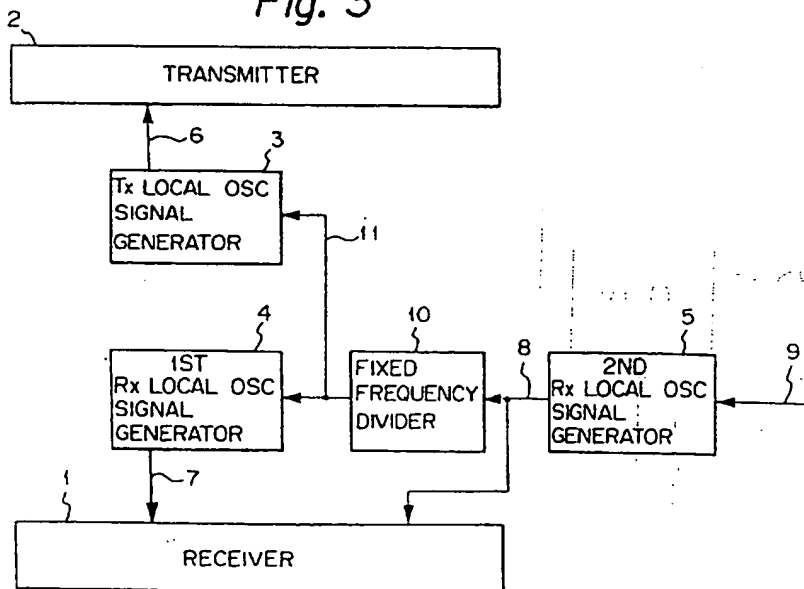
100 Grays Inn Road, LONDON, WC1X 8AL,

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(54) Transmitter and/or superheterodyne receiver

(57) A transmitter and/or superheterodyne receiver controlled by an externally-supplied reference frequency signal, including signal-conditioning means for removing noise from the reference signal comprising a low loop gain phase locked loop circuit 5 to produce an output having a frequency which is a multiple of that of the reference signal and a fixed frequency divider 10 for dividing said output to produce a conditioned reference signal of a frequency equal to, or having a fixed relationship to, that of the externally-supplied reference signal. The phase locked loop may include a VCO, a phase comparator, 2 low pass filter and a variable frequency divider.

Fig. 3



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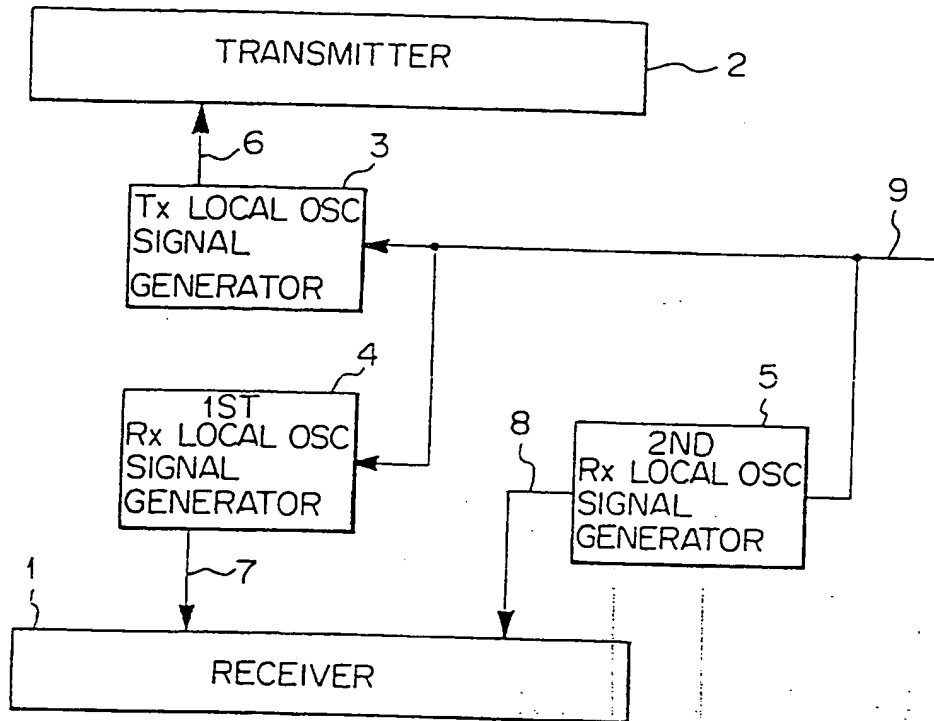
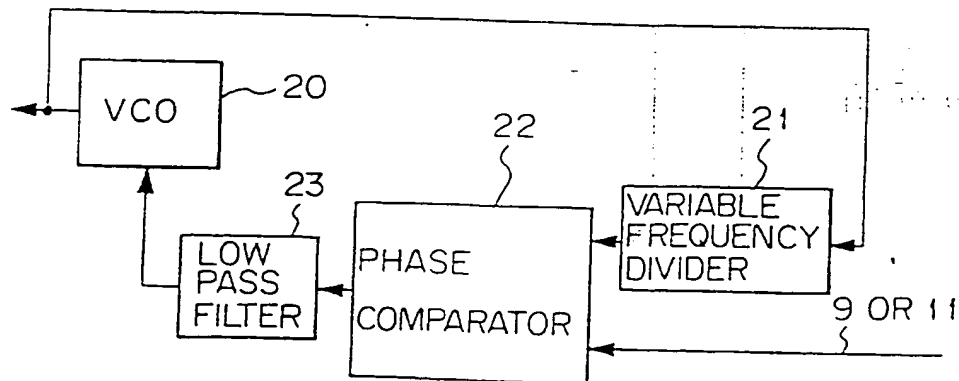
*Fig. 1* PRIOR ART*Fig. 2* PRIOR ART

Fig. 3

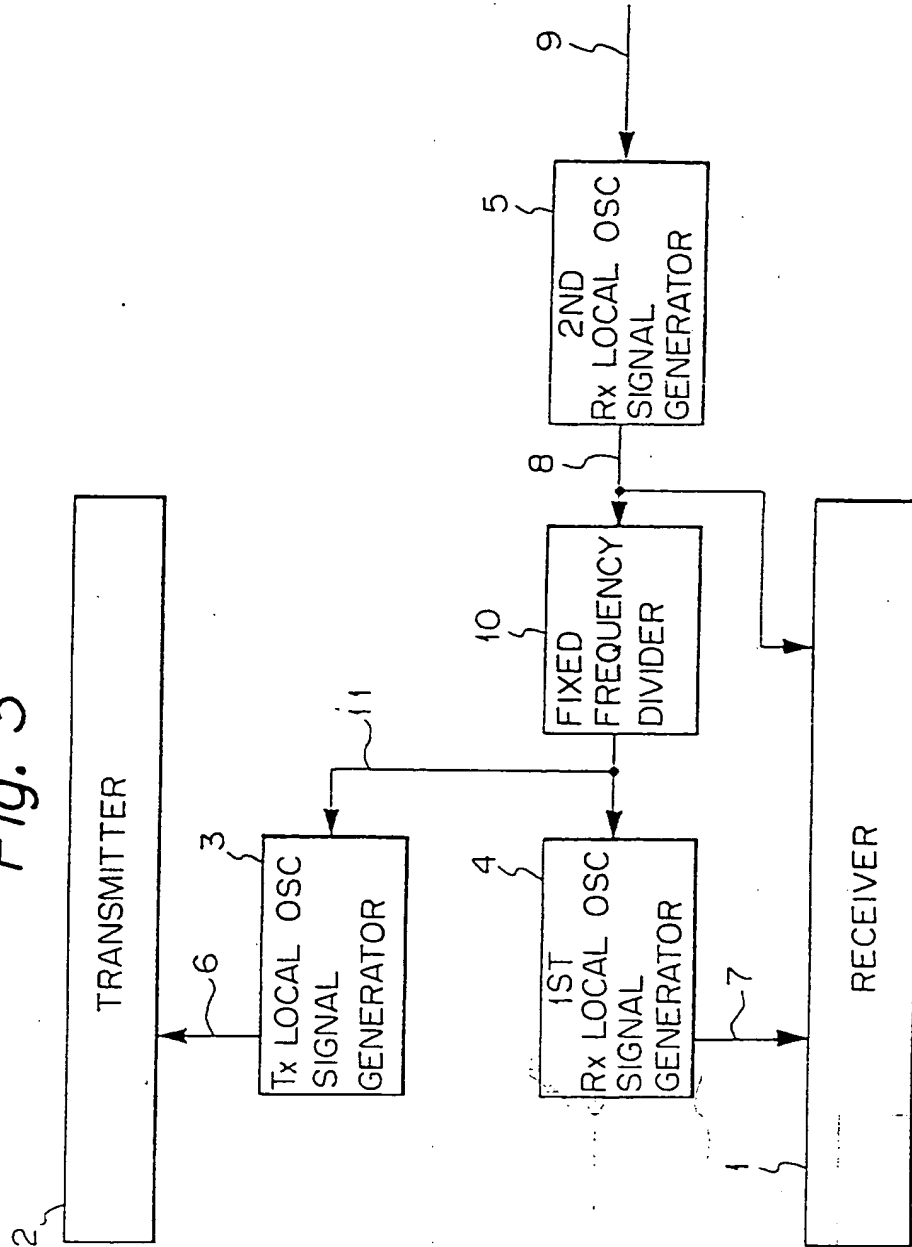


Fig. 4A

Fig. 4

Fig. 4A	Fig. 4B	Fig. 4C
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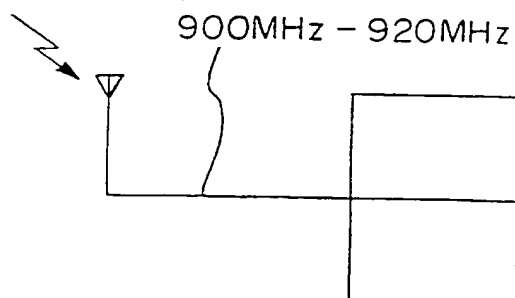


Fig. 4B

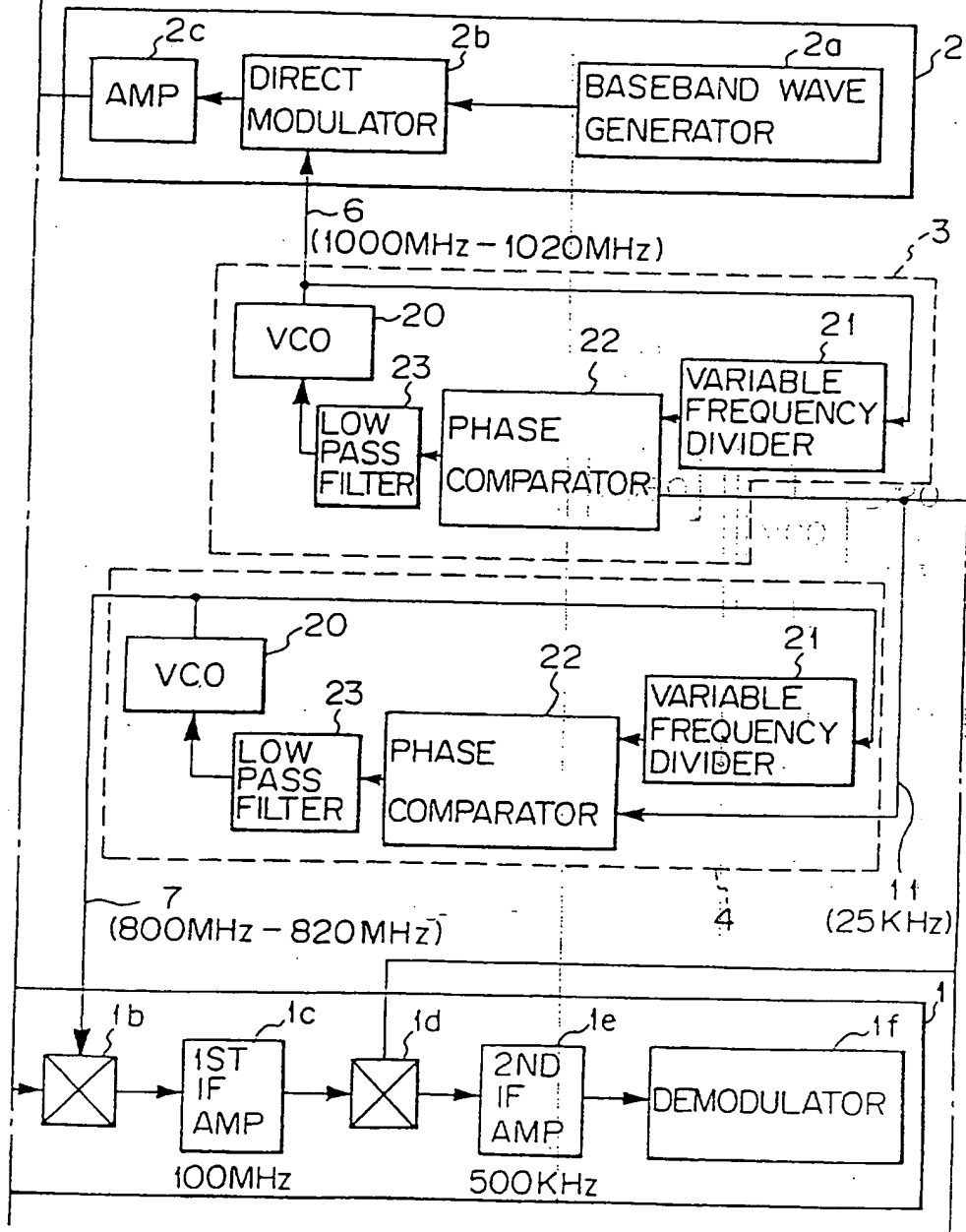
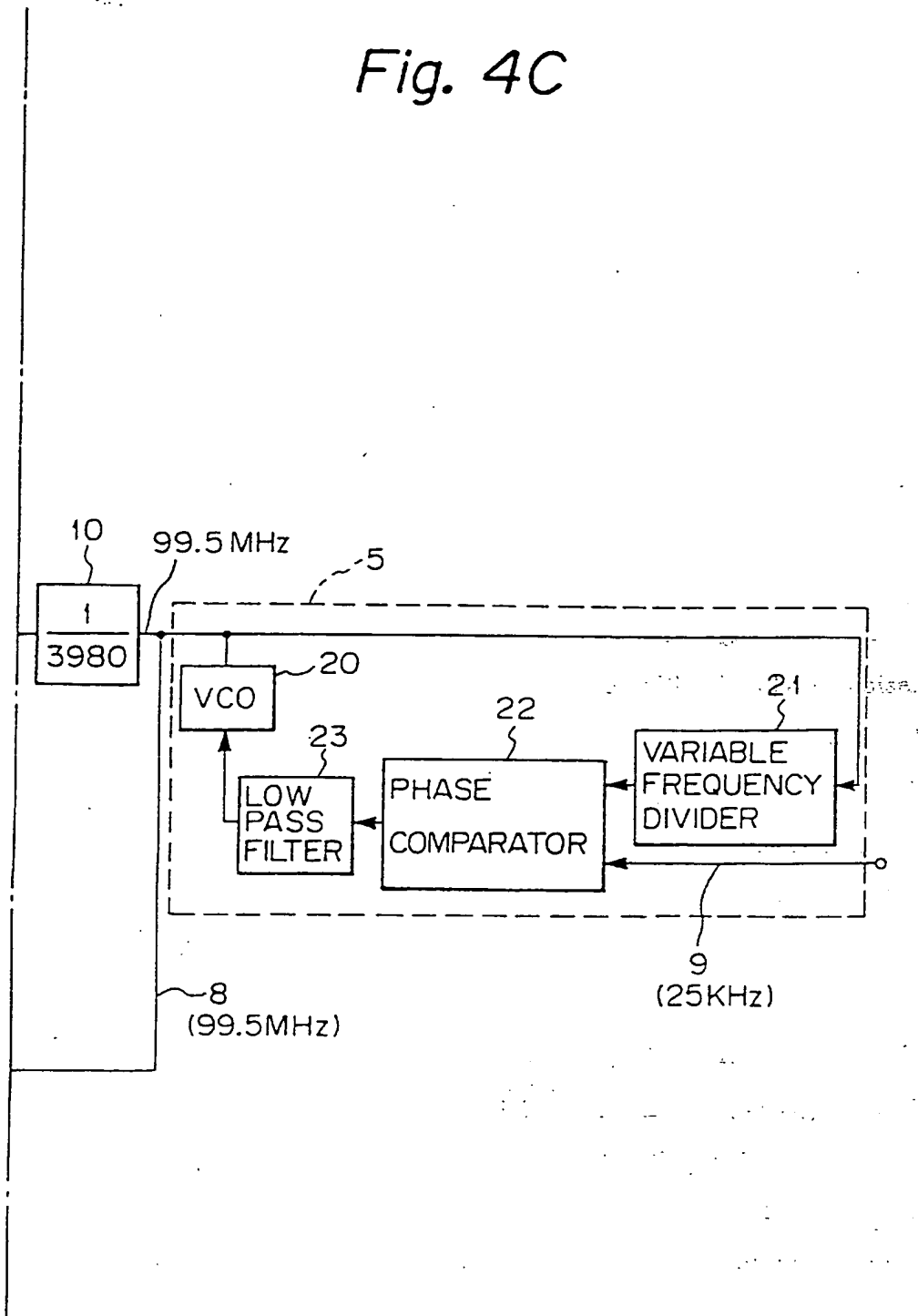


Fig. 4C



## TRANSMITTER AND/OR SUPERHETERODYNE RECEIVER

## BACKGROUND OF THE INVENTION

The present invention relates to a transmitter and/or superheterodyne receiver and, more particularly, to a transmitter/receiver which prevents the modulation accuracy of a signal to be transmitted and reception sensitivity from being degraded even when a reference signal fed from the outside involves noise.

There has been known a transmitter/receiver made up of a superheterodyne receiver, a transmitter, a local oscillation signal generating section for transmission, and a first and a second local oscillation signal generating section for reception. The local oscillation signal generating sections are each implemented as a phase locked loop (PLL) frequency synthesizer. The PLL synthesizer has a voltage controlled oscillator (VCO), a variable frequency divider for sequentially dividing the output frequencies of the VCO, a phase comparator for comparing the phase of the output of the frequency divider with that of a reference signal, and a low pass filter for smoothing the output of the phase comparator to thereby produce a control signal. Controlled by this control

signal, the VCO outputs a signal synchronous to the reference  
signal. The transmission and first reception local oscillation  
signal generating sections are each implemented by a PLL  
capable of switching the frequency rapidly, i.e., having a high  
5 loop gain since it has to change the frequency in association  
with a transmission frequency or a reception frequency.  
Conversely, the second reception local oscillation signal  
generating section is implemented by a PLL having a low loop  
gain since the frequency thereof needs only to be constant.

10 The conventional transmitter/receiver has a problem  
left unsolved, as follows. When the reference signal from the  
outside involves noise, the transmission and first reception  
local oscillation signal generating sections cannot remove the  
noise due to their high loop gains. As a result, the  
15 transmission and first reception local oscillation signals  
from such sections are unstable and low in carrier-to-noise  
(C/N) characteristic, degrading the modulation accuracy of a  
signal to be transmitted as well as a reception  
characteristic.

20

It is, therefore, an object of at least the preferred  
embodiment of the present invention to  
provide a transmitter/receiver capable of preventing the  
stability of the frequencies of local oscillation signals for  
25 transmission and reception and C/N characteristic from being



degraded even when a reference signal introduced from the outside involves noise.

Accordingly, the present invention provides a transmitter and/or superheterodyne receiver controlled by an externally-supplied reference frequency  
5 signal, including signal-conditioning means for removing noise from the reference signal comprising a low loop gain phase locked loop circuit to produce an output having a frequency which is a multiple of that of the reference signal and a fixed frequency divider for dividing said output to produce a conditioned reference signal of a frequency equal to, or having a fixed relationship to, that of the externally-  
10 supplied reference signal.

According to the above, the transmitter and/or superheterodyne receiver can prevent the modulation accuracy of a signal to be transmitted and reception sensitivity from being degraded even when the reference signal introduced from the outside includes noise.

15 Other preferred features of the invention are as set out in the subordinate claims, which are deemed repeated here as consistory clauses.

The above and other objects, features and advantages of the present invention will become more apparent from the

following detailed description taken with the accompanying drawings in which:

FIG. 1 is a block diagram schematically showing a conventional transmitter/receiver having a superheterodyne receiver;

FIG. 2 is a schematic block diagram of a PLL frequency synthesizer included in the transmitter/receiver of FIG. 1 and implementing a local oscillation signal generating section;

FIG. 3 is a schematic block diagram showing a transmitter/receiver embodying the present invention; and

FIG. 4 is a block diagram showing the embodiment more specifically.

To better understand the present invention, a brief reference will be made to a conventional transmitter/receiver, shown in FIG. 1. As shown, the transmitter/receiver has a superheterodyne receiver 1, a transmitter 2, a transmit (Tx) local oscillation signal generating section 3, a first receive (Rx) local oscillation signal generating section 4, and a second Rx local oscillation signal generating section 5. A reference signal 9 is fed to the local oscillation signal generating sections 3, 4 and 5 from the outside of the transmitter/receiver. Among the local oscillation signal generating sections 3-5, the section 3

generates a Tx local oscillation signal 6 and applies it to the transmitter 2 for converting a baseband signal to be transmitted to an RF signal. The section 4 generates a first Rx local oscillation signal 7 and delivers it to the receiver 1 for converting a received RF signal to a first IF signal. Further, the section 5 generates a second Rx local oscillation signal 8 and feeds it to the receiver 1 for transforming the first IF signal to a second IF signal. As shown in FIG. 2, these sections 3, 4 and 5 are each implemented as a PLL frequency synthesizer. The PLL synthesizer is made up of a VCO 20, a variable frequency divider 21 for sequentially dividing the output frequencies of the VCO 20, a phase comparator 22 for comparing the phase of the output of the frequency divider 21 with that of the reference signal 9, and a low pass filter 23 for smoothing the output of the phase comparator 22 to thereby produce a control signal. Controlled by this control signal, the VCO 20 outputs a signal synchronous to the reference signal 9. The local oscillation signal generating sections 3 and 4 are each implemented by a PLL capable of switching the frequency rapidly, i.e., having a high loop gain since it has to change the frequency in association with a transmission frequency or a reception frequency. Conversely, the local oscillation signal generating section 5 is implemented by a PLL having a low loop gain since the frequency thereof needs only to be constant. The reference

signal 9 is fed to the phase comparator 22 from the outside of the transmitter/receiver; the local oscillation signal is synchronous to the reference signal 9.

5 The problem with the conventional transmitter/receiver is that when the reference signal involves noise, the local oscillation signal generating sections 3 and 4 cannot remove the noise due to their high loop gains. As a result, the local oscillation signals 6 and 7 from these sections 3 and 4 are unstable and low in C/N characteristic, degrading the  
10 modulation accuracy of a signal to be transmitted as well as a reception characteristic.

Referring to FIGS. 3 and 4 a transmitter/receiver embodying the present invention will be described. In the figures, the same or similar constituent parts as or to the  
15 parts shown in FIGS. 1 and 2 are designated by the same reference numerals, and a detailed description thereof will not be made in order to avoid redundancy. As shown, in the illustrative embodiment, a reference signal 9 from the outside is fed only to a second Rx local oscillation signal  
20 generating section 5. The output of the second Rx local oscillation signal generating section 5 branches off to a fixed frequency divider 10 and a receiver 1. The fixed frequency divider 10 has an output 11 thereof connected to a Tx local oscillation signal generating section 3 and a first Rx local  
25 oscillation signal generating section 4. In this configuration,

the second Rx local oscillation signal generating section 5 generates a second local oscillation signal 8 for reception. This signal 8 has the frequency thereof divided by the fixed frequency divider 10. The resulting output 11 of the frequency divider 10 plays the role of a reference signal  
5 meant for the local oscillation signals 3 and 4.

For each of the local oscillation signal generating sections 3, 4 and 5, use is also made of the PLL frequency synthesizer shown in FIG. 2. Specifically, the reference  
10 signal 9 from the outside is input to the phase comparator 22 of the second Rx local oscillation signal generating section 5. Since the second Rx local oscillation signal 8 for reception should only have a fixed frequency, the local oscillation signal generating section 5 may generally have an extremely  
15 low loop gain. Hence, this section, or PLL, 5 produces an output free from the influence of noise which is entrained by the reference signal 9.

Assume that the reference signal 9 has a frequency  $f_{REF}$ , and that the variable frequency divider 21, FIG. 2, has a denominator N. Then, the frequency  $f_{R2}$  of the second local  
20 oscillation signal 8 for reception is expressed as:

$$f_{R2} = f_{REF} \times N$$

The signal 8 is applied to the receiver 1 on one hand and to the fixed frequency divider 10 on the other hand. Assuming that the fixed frequency divider 10 has a denominator N, the output 11 of the frequency divider 10 has a frequency  $f_N$  produced by:

$$f_N = f_{R2} / N = f_{REF}$$

It follows that the signal 11 has the same frequency as the reference signal 9 and has had the noise removed by the second Rx local oscillation signal generating section, or PLL, 5.

The signal 11 is applied to the phase comparator 22, FIG. 2, included in each of the local oscillation signal generating sections 3 and 4 as a reference signal. The local oscillation signal generating sections 3 and 4 are each implemented by a PLL having a high loop gain since it has to switch the frequency rapidly in association with the frequency for transmission or reception. However, since the reference signal 11 applied to these sections 3 and 4 is free from noise components, the local oscillation signals 6 and 7 from the sections 3 and 4 are stable in frequency and desirable in C/N characteristic. The denominator N of the fixed frequency divider 10 may be 2N in place of N, in which case the signals

6 and 7 will each have a frequency whose minimum variable interval is  $f_{REF}/2$ .

Referring to FIG. 4, a more specific construction of the embodiment is shown. The transmitter/receiver is assume to  
5 have a reception frequency ranging from 900 MHz to 920 MHz and a transmission frequency ranging from 1000 MHz to 1020 MHz and have channel intervals of 25 kHz. As shown in FIG. 4, the receiver 1 has an RF amplifier 1a, a first mixer 1b, a first IF amplifier 1c, a second mixer 1d, a second IF amplifier 1e,  
10 and a demodulator 1f. The transmitter 2 has a baseband waveform generator 2a, a direct modulator 2b, and an amplifier 2c. The first and second IF frequencies are 100 MHz and 500 kHz, respectively. The Tx local oscillation signal 6, first Rx local oscillation signal 7 and second Rx local  
15 oscillation signal 8 respectively have frequencies of 1000 MHz to 1020 MHz, 800 MHz to 820 MHz, and 99.5 MHz. Since the second Rx local oscillation signal generating section 5 needs only a fixed oscillation frequency, the VCO 20 thereof is provided with control voltage sensitivity of as low as  
20 1 ppm/V; the PLL also has an extremely low loop gain. By contrast, in each of the Tx local oscillation signal generating section 3 and first Rx local oscillation signal generating section 4, the VCO 20 is provided with control voltage sensitivity of about 10 MHz in order to cover a 20 MHz band;  
25 the PLL also has a loop gain high enough to effect rapid

frequency switching. The reference signal 9 from the outside is applied to the second Rx local oscillation signal generating section 5. The 99.5 MHz second Rx local oscillation signal 8 from the section 5 is applied to both the second mixer 1d of the receiver 1 and the fixed frequency divider 10 having a denominator N of 3980.

In the above condition, the output signal 11 of the fixed frequency divider 10 has a frequency  $f_N$  produced by:

10             $f_N = 99.5 \text{ (MHz)} / 3980 = 25 \text{ (kHz)}$

The signal 11 is fed to the Tx local oscillation signal generating section 3 and first Rx local oscillation signal generating section 4. As a result, the frequency of the Tx local oscillation signal 6 is variable over a range of from 1000 MHz to 1020 MHz at intervals of 25 kHz in association with the transmission frequency. Likewise, the frequency of the Rx local oscillation signal 7 is variable over a range of from 800 MHz to 820 MHz at intervals of 25 kHz in association with the reception frequency. The signals 6 and 7 are respectively applied to the direct modulator 2b of the transmitter 2 and the first mixer 1b of the receiver 1.

In summary, it will be seen that the present invention provides a transmitter/receiver which generates, even when a reference signal introduced from the outside involves noise,



local oscillation signals which are stable and desirable in C/N characteristic. Such local oscillation signals prevent modulation accuracy for transmission and reception sensitivity from being degraded. This advantage is derived  
5 from a unique arrangement wherein the reference signal from the outside is fed only to a second reception local oscillation signal generating section whose loop gain is low, while a reference signal for the other local oscillation signal generating sections is produced by dividing the frequency of a  
10 second reception local oscillation signal.

Various modifications will become possible for those skilled in the art after receiving the teachings of the present disclosure without departing from the scope thereof.

CLAIMS

1. A transmitter and/or superheterodyne receiver controlled by an externally-supplied reference frequency signal, including signal-conditioning means for removing noise from the reference signal comprising a low loop gain phase locked loop circuit to produce an output having a frequency which is a multiple of that of the reference signal and a fixed frequency divider for dividing said output to produce a conditioned reference signal of a frequency equal to, or having a fixed relationship to, that of the externally-supplied reference signal.
2. A transmitter and/or superheterodyne receiver according to Claim 1, including a superheterodyne receiver, a transmitter, a transmission local oscillation signal generating means for receiving said conditioned reference signal and outputting a transmission local oscillation signal to said transmitter, and a reception local oscillation signal generating means for receiving said conditioned reference signal and outputting a reception local oscillation signal to said receiver.
3. A transmitter and/or superheterodyne receiver according to Claim 2, wherein each of said generating means comprises a high loop gain phase locked loop circuit.
4. A transmitter and/or superheterodyne receiver according to Claim 3, wherein each phase locked loop circuit comprises a voltage controlled oscillator (VCO), a variable frequency divider for sequentially dividing output frequencies of the VCO, a phase comparator for comparing a phase of an output of said divider and a phase of the reference signal received by the circuit, and a low pass filter for smoothing an output of said phase comparator to produce a control signal and feeding said control signal to the VCO.
5. A transmitter and/or superheterodyne receiver as claimed in any preceding claim and substantially as herein described with reference to and as shown in Figure 3 or Figure 4 of the accompanying drawings.

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